Waveguide Cavity Resonator



The cavity resonator is obtained from a section of rectangular wave guide, closed by two additional metal plates. We assume again perfectly conducting walls and loss-less dielectric.

The addition of a new set of plates introduces a condition for standing waves in the z-direction which leads to the definition of oscillation frequencies

$$f_{c} = \frac{1}{2\sqrt{\mu\varepsilon}} \sqrt{\left(\frac{m}{a}\right)^{2} + \left(\frac{n}{b}\right)^{2} + \left(\frac{l}{d}\right)^{2}}$$

The high-pass behavior of the rectangular wave guide is modified into a <u>very narrow</u> pass-band behavior, since cut–off frequencies of the wave guide are transformed into oscillation frequencies of the resonator.



The cavity resonator will have modes indicated as

TE_{mnl} TM_{mnl}

The value of the index corresponds to periodicity (number of half sine or cosine waves) in the three directions. Using z-direction as the reference for the definition of transverse electric or magnetic fields, the allowed indices are

 $TE \begin{cases} m = 0, 1, 2, 3... \\ n = 0, 1, 2, 3... \\ l = 1, 2, 3... \\ m = 1, 2, 3... \\ l = 0, 1, 2, 3... \\ l = 0, 1, 2, 3... \\ m \text{ or } n \text{ allowed} \end{cases} m = 1, 2, 3... \\ l = 0, 1, 2, 3... \\ l$

The mode with **lowest** resonance frequency is called **dominant** mode. In the case $a \ge d > b$ the dominant mode is the TE₁₀₁.

Note that for a TM cavity mode, with magnetic field transverse to the z-direction, it is **possible** to have the third index equal to zero. This is because the magnetic field is going to be parallel to the third set of plates, and it can therefore be uniform in the third direction, with no periodicity.

The electric field components will have the following form that satisfies the boundary conditions for perfectly conducting walls

$$E_{x} = E_{ox} \cos\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) \sin\left(\frac{l\pi}{d}z\right)$$
$$E_{y} = E_{oy} \sin\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) \sin\left(\frac{l\pi}{d}z\right)$$
$$E_{z} = E_{oz} \sin\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) \cos\left(\frac{l\pi}{d}z\right)$$

The magnetic field intensities are obtained from Ampere's law

$$H_{x} = \frac{\beta_{z}E_{y} - \beta_{y}E_{z}}{j\omega\mu} \sin\left(\frac{m\pi}{a}x\right)\cos\left(\frac{n\pi}{b}y\right)\cos\left(\frac{l\pi}{d}z\right)$$
$$H_{y} = \frac{\beta_{x}E_{z} - \beta_{z}E_{x}}{j\omega\mu}\cos\left(\frac{m\pi}{a}x\right)\sin\left(\frac{n\pi}{b}y\right)\cos\left(\frac{l\pi}{d}z\right)$$
$$H_{z} = \frac{\beta_{y}E_{x} - \beta_{x}E_{y}}{j\omega\mu}\cos\left(\frac{m\pi}{a}x\right)\cos\left(\frac{n\pi}{b}y\right)\sin\left(\frac{l\pi}{d}z\right)$$

Similar considerations for modes and indices can be made if the other axes are used as reference for transverse fields, leading to analogous resonant field configurations.

A cavity resonator can be coupled to a wave guide through a small opening. When the input frequency resonates with the cavity, electromagnetic radiation enters the resonator and a lowering in the output is detected. With carefully tuned cavities, this scheme can be used for frequency measurements.



Examples of resonant cavity excited by using coaxial cables.



The termination of the inner conductor of the cable acts like an elementary dipole (left) or an elementary loop (right) antenna.